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**SUBJECT: RESISTIVE WALL MODE SYSTEM SCOPING STUDY**

**Reference:** 72-030123-JM-01 “Preliminary Requirements For NSTX Resistive Wall Mode And Error Field Feedback Coils And Power Supplies”, J. Menard, S. Sabbagh

### Summary

The purpose of this memo is to 1) identify the design issues, 2) develop the design concepts and 3) outline the tasks related to the implementation of a Resistive Wall Mode (RWM) and error field correction system on NSTX.

Since a GA/PPPL collaboration has already deployed an RWM system on DIIID it is advantageous and cost effective for NSTX to build upon this experience and utilize coil and power supply concepts already developed. Therefore a starting point for this study is to determine to what extent the GA coil and power supply designs can be adopted for NSTX duty.

It is concluded herein that the DIIID internal coil design could be adopted on NSTX, and that the Switching Power Amplifiers (SPAs) are suitable.

A decision on internal or external coil location needs to be made soon, based upon a trade-off between costs and benefits. Toward this end, the next steps are to 1) quantify the performance difference (VALEN code) and 2) quantify the cost/schedule difference. The results given herein should provide the necessary input to these steps.

### Requirements

Per the reference memo and discussions with J. Menard et al the proposed NSTX system requirements are as follows.

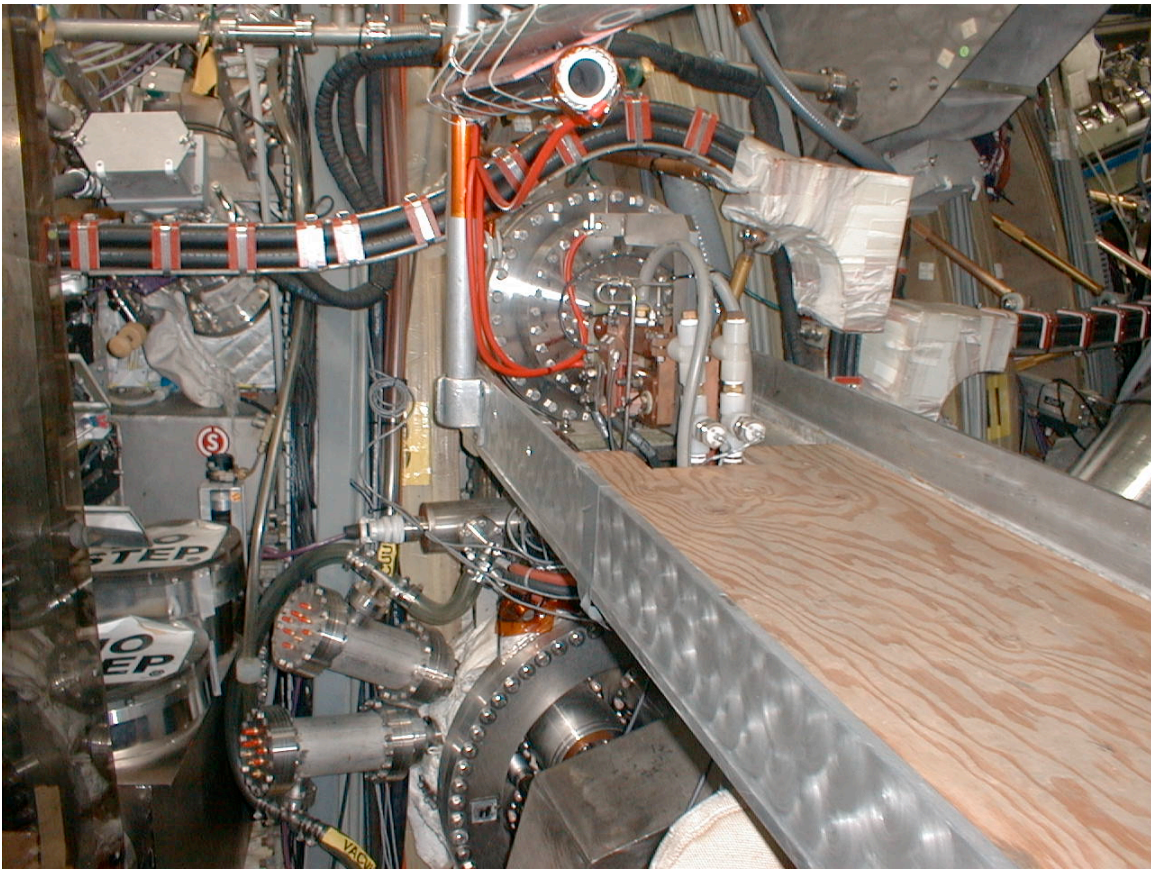
Number of Coils	6
Coil Location	Centered about midplane
Coil Connections	Diametrically opposite coils connected in anti-series forming 3 independently controllable circuits
Coil height	Approximately equal to gap between passive plates ~ 1 m.
DC Field at $r=R_0+0.6*a$	50 gauss

AC Field at 1kHz, $r=R0+0.6*a$	10 gauss
Maximum Ripple	+/- 2% of full load DC
Pulse Length	5 sec
Repetition Period	300 sec

## GA RWM System

### *Internal and External Coils*

GA started their RWM experimental program with a coil system consisting of a set of 6 external coils centered about the midplane<sup>1</sup>. The coils were wound using 4 turns of 750MCM cable carrying up to 5kA for 10s. These were designed to produce 30 gauss at the plasma edge. They were located at approximately  $R=2.56m$ , or 0.22m from the plasma edge. See figure 1.

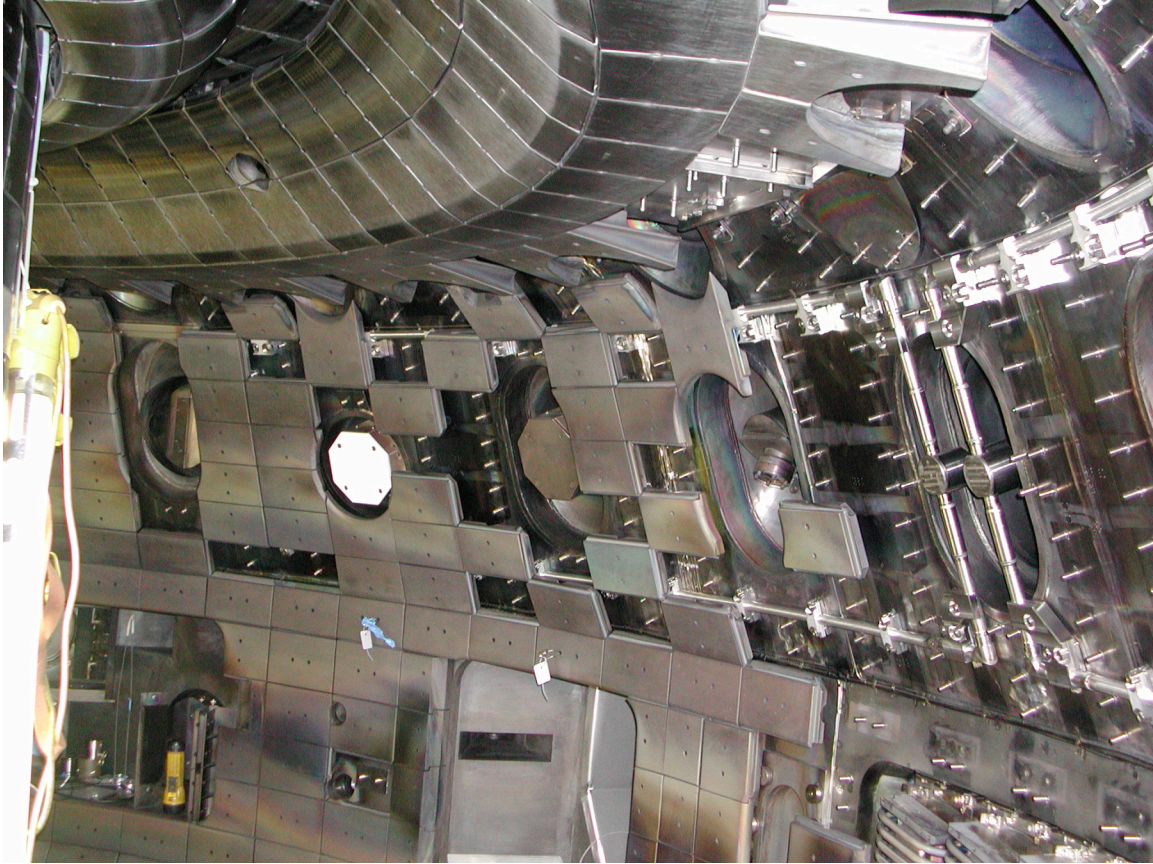


*Figure 1 – DIII-D External Coils*

More recently an internal coil system has been added which consists of 12 coils, 6 above and 6 below midplane<sup>2</sup>. See figure 2.

<sup>1</sup> “Resistive Wall Mode Feedback System on DIII-D”, J. Scoville et al, 18<sup>th</sup> SOFE, Albuquerque, Oct. 1999

<sup>2</sup> “Design, Fabrication, Installation and Testing of In-Vessel Control Coils for DIII-D”, P. Anderson et al, 22<sup>nd</sup> SOFT, Helsinki, Sept 2002



*Figure 2 – DIII-D Internal Coils*

The coils are approx. 0.5 m tall by 2.0 meter wide, single turn water cooled tubular copper conductors (14.4mm OD, 8.6mm ID) wrapped with one half-lapped layer of polyamide (0.5mm (0.002”) thickness Kapton) insulation, a 1.7 mm thick Vespel spacer, and an additional layer of Kapton tape, all encased in a 304SS tube (19mm OD). The coils are capable of 7kA continuous with 5.5m/sec water cooling. Because a fully assembled coil cannot be passed into the vessel, the coil is fabricated outside the vessel in three pieces; a lower conductor assembly, an upper conductor assembly and a 35 mm diameter concentric lead assembly. Leads are brought out from two coils at a time through 10” diameter openings.

The design requires in-vessel joining of copper with three induction brazes and six orbital welds of the stainless vacuum jacket. The coils are installed under PFC tiles and are baked to 350°C. The design provides a double barrier of copper and stainless steel against water leakage into the machine. In order to detect water leaks in the copper or leaks through the stainless into the vessel, the insulation space is sealed in two places outside the vessel using machined polyamide and “O” rings. During the initial bake of the vessel, this space is vacuum pumped to remove moisture and gases that evolve. After cool down, dry nitrogen gas is back filled to about 0.7 bar, and sealed off. The pressure in this trapped volume is monitored to detect either water leaks through the copper (higher pressure) or stainless sheath leaks to the vacuum (lower pressure). During vessel baking,

the cooling water in the coils is replaced with dry nitrogen in order to limit oxidization of the copper.

An extensive R&D program was conducted in '00 and '01 which led to the installation of two prototype coils in DIIID. After their removal they were hipot tested (to destruction) and found to withstand 5kV.

The 12 production coils were fabricated during the first 6 months of '02 and then installed during a 3 month opening. So far there have been no leaks or any other problems with the coils.

The total cost associated with the fabrication and installation of the 12 production coils was \$2M, most of which was for labor costs. However this figure includes costs for a significant amount of PFC tile removal, modification, and re-installation, along with the relocation and replacement of some magnetic diagnostics. So the cost associated with the coils themselves was perhaps of the order of \$1.5-\$1.8M or approx. \$100K-\$150K per coil.

If NSTX elects to install internal coils, the GA design could be adapted, taking advantage of their prior R&D and experience. Furthermore the coils could be fabricated by GA and they could lead in the installation task, especially the in-vessel brazing and welding operation. In fact the NSTX installation would be simpler and less demanding due to the lower bakeout requirement (the coils would be attached to the vessel wall at 150°C), simpler geometry (the coils on the midplane would conform to the cylindrical shape of the vessel), and lower fields (and forces). It may be possible to install the coils through Bay K on NSTX (future NBI port) and avoid much of the in-vessel fabrication, which would be a great simplification. Similarly, if the coils were installed during a center stack removal then would be possible to pass them into the vessel via the opening at the top of NSTX. One additional challenge presented by NSTX is the CHI operation which places a 1kV bias on center stack casing. But part of the voltage can appear on the resistive grounded outer VV. So RWM coils mounted on the outer VV of NSTX should be designed for a hipot of  $2*1+1=3\text{kV}$ . GA tests of their insulation scheme, to destruction, indicate a dielectric strength of 5kV.

### *Power Supplies*

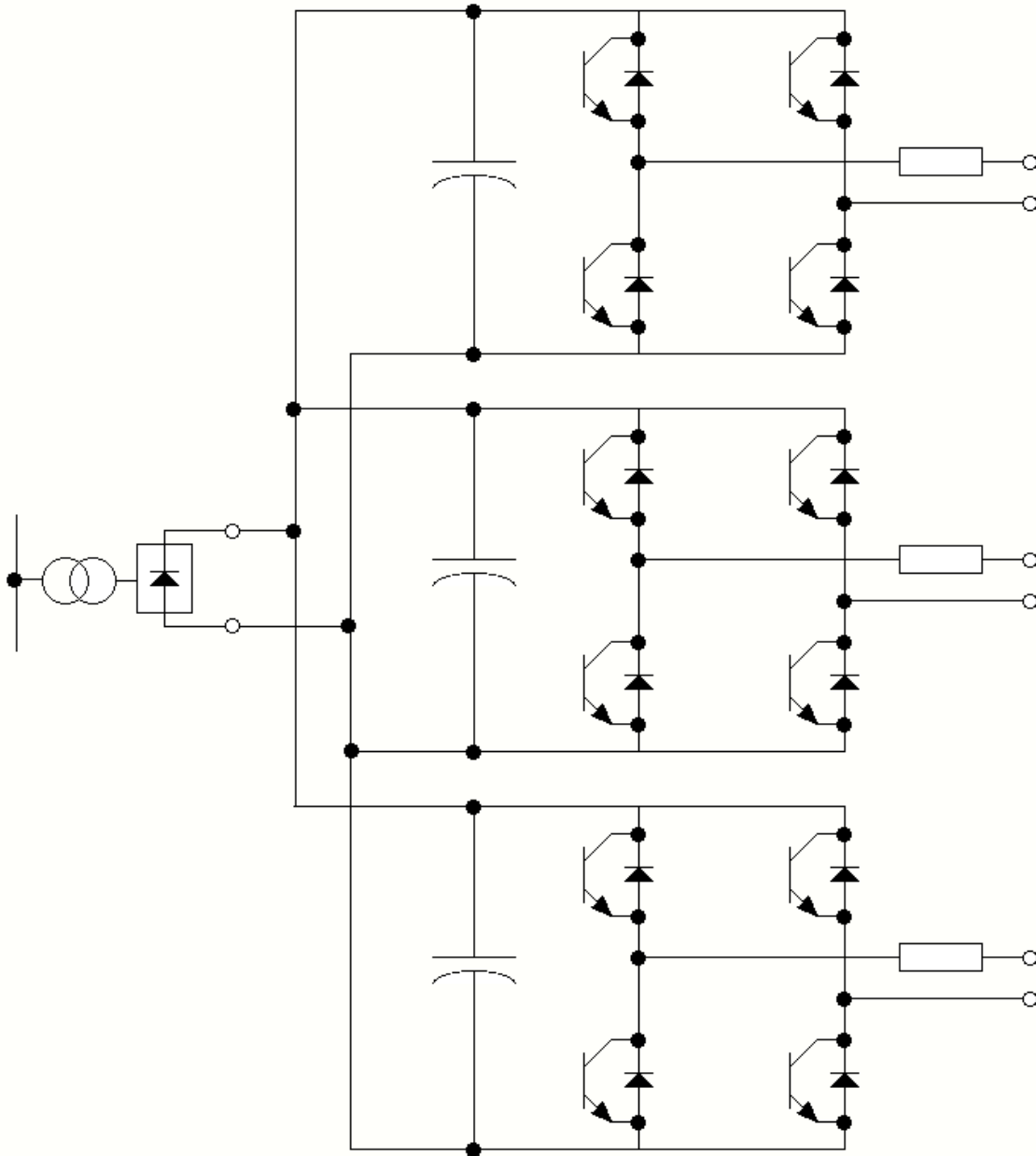
The power supplies consist of two parts, a rectifier DC source and an IGBT chopper "Switching Power Amplifier" (SPA)<sup>3</sup>. One DC source can be connected to supply one or more SPA units.

The SPA choppers are subdivided into three parallel modules which can be controlled separately and connected to three independent loads or combined in parallel to drive one load. The DC inputs to the three chopper modules are connected in parallel, and each to

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<sup>3</sup> Operating & Maintenance Manual, PU PO#S-04108-G, Robicon SO#1-64739

chopper module input includes an input filter capacitor and output filter inductor. See figure 3.



*Figure 3 – DIID DC Supply and SPA Configuration*

The SPAs were specified and procured by PPPL and manufactured by Robicon. The DC source supplies were obtained from LLNL surplus. SPA characteristics are given in the following table.

Pulse Current per Module	1.667kA
Pulse Current (3 parallel module)	5kA
Pulse Duration	10 sec
Pulse Period	180 sec
Input Voltage	300Vdc
Switching Frequency	3.5kHz (first version) 7.0kHz (second version)
Input Filter Capacitance (3 parallel module)	0.2835 millifarad, 800V
Output Filter Inductor (per module)	11 microhenry
Output Filter Inductor (3 parallel module)	3.67 microhenry

The SPA IGBTs consist, in each bridge arm, 2 parallel IGBT devices (EUPEC 400A/1200V).

If NSTX elects to procure the same SPA units then there is a possibility that they could be operated at a somewhat higher voltage. For the DC source, therefore, it may be possible to utilize PPPL Transrex rectifiers which at present produce a DC voltage of 1012.85Vdc at  $\alpha=0$ . It would be necessary to either supply the Transrex units with a lower AC input voltage (480Vac would yield 650Vdc) or control them to a lower voltage with suitable accessories to prevent overvoltage. Alternatively, new DC source power supplies could be procured or obtained from surplus TBD.

#### Analysis of Implementation Options

A spreadsheet analysis was performed for several NSTX design options. This was helpful in identifying the issues. The GA internal and external coil designs were also analyzed so as to provide a benchmark.

The complexities of the magnetics (mutual coupling to VV and resultant eddy current effects) were not included in the model. However, skin effect was accounted for. In arriving at the results in the GA cases, the physical dimensions, circuit resistances and inductances, including choke values, were based on information extracted from several GA publications<sup>1,2,4</sup>. In the NSTX cases the choke value was chosen so as to limit the peak-to-peak ripple to 4% of the total amp-turns.

The following table provides a summary of the cases studied.

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<sup>4</sup> "Modeling and Design of A Resistive Wall Mode Stabilization System with Internal Field Coils In DIII-D", G. Jackson et al, 44<sup>th</sup> APS, Orlando, Nov. 2002

	GA Internal Coil	GA External Coil	NSTX Internal Coil	NSTX External Coil	NSTX External Coil
#Turns	1	4	1	1	2
Peak DC Current	5kA/5kA-turn	5kA/20kA-turn	5kA/5kA-turn	5kA/5kA-turn	5kA/10kA-turn
Rcoil	R0+a=2.34m (see note 1)	2.56m	1.68m (5/8" inside VV wall)	1.85m (6" outside VV wall)	1.85m
dZcoil	1.02m	1.6m	1.0m	1.0m	1.0m
Rpoint (target for field)	R0+0.6*a=2.07m	2.07m	R0+0.6*a=1.26m	1.26m	1.26m
Br @ Rpoint (see note 2)	9.6e-3 gauss/amp-turn	6.65e-3 gauss/amp-turn	9.05e-3 gauss/amp-turn	6.98e-3 gauss/amp-turn	6.98e-3 gauss/amp-turn
Br @ Rpoint @ Peak DC Current	48.1 gauss	133.0 gauss	45.2 gauss	34.9 gauss	69.8 gauss
Br @ Rpoint @ 1kHz	4.8 gauss	1.1 gauss	6.3 gauss	4.9 gauss	4.9 gauss
Coil Conductor	GA	750MCM cable	GA	500MCM cable	250MCM cable
Coil Impedance	Rdc=1.33mΩ L=6.6μH	Rdc=1.64mΩ L=128μH	Rdc=1.1mΩ L=5.1μH	Rdc=0.54mΩ L=5.25μH	Rdc=1.86mΩ L=22.6μH
#Series Coils	2	2	2	2	2
DC Feed Cable	Rdc=6.33mΩ L=32.8μH	Rdc=6.33mΩ L=32.8μH	2x250' 500MCM cable Rdc=35.1mΩ L=56.3μH	2x250' 500MCM cable Rdc=35.1mΩ L=56.3μH	2x250' 500MCM cable Rdc=35.1mΩ L=56.3μH
SPA Switching Frequency	3.5kHz	3.5kHz	7.0kHz	7.0kHz	7.0kHz
SPA Voltage	300V	300V	300V	300V	300V
Lchoke	50μH	0	1.4μH	1.2μH	0
Peak-peak ripple (see note 3)	5.7%/2.8gauss	1.9%/2.5gauss	4.0%/1.8gauss	4.0%/1.4gauss	4.0%/2.8gauss

Notes:

- 1) GA Internal coils consist of 2 sets, one above and one below mid. This radius chosen to simulate such coils using 1 set centered around the midplane.
- 2) Br estimated based on planar, anti-series coil pair
- 3) % based on amp-turns

The following curves depict the dependence of current, field, and pulse length on frequency for the various case, considering the variation in inductive impedance as well as the skin effect on resistance.

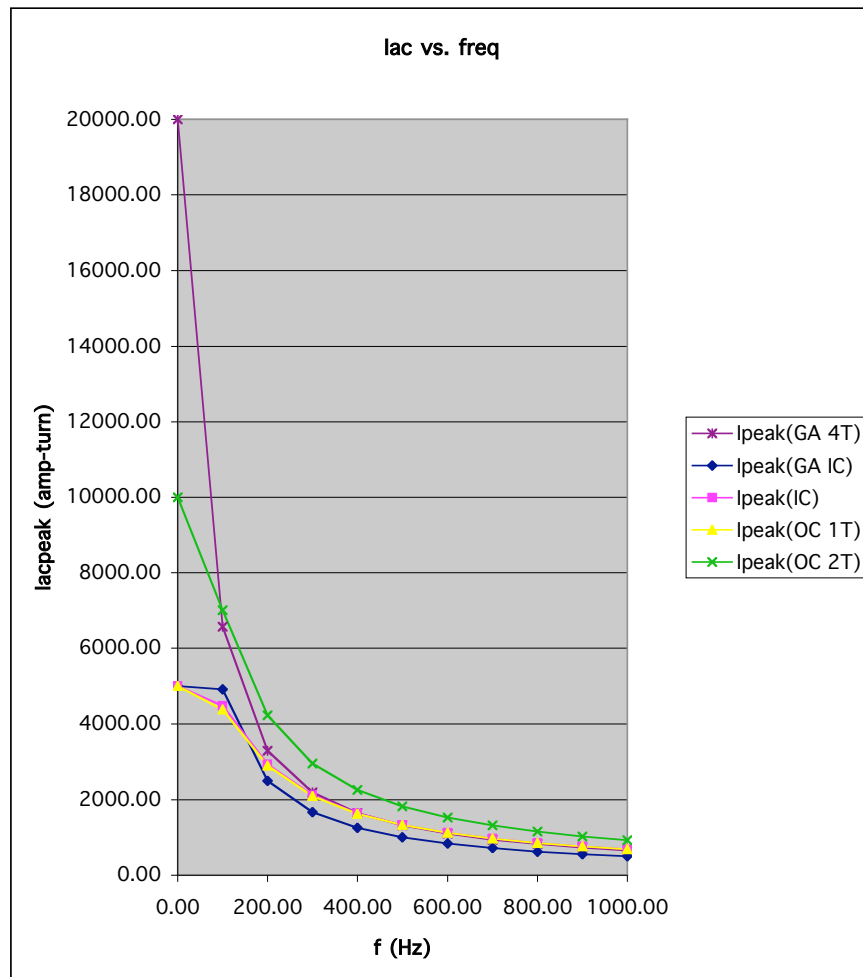


Figure 4 – Available Peak Current (amp-turns) vs. Frequency



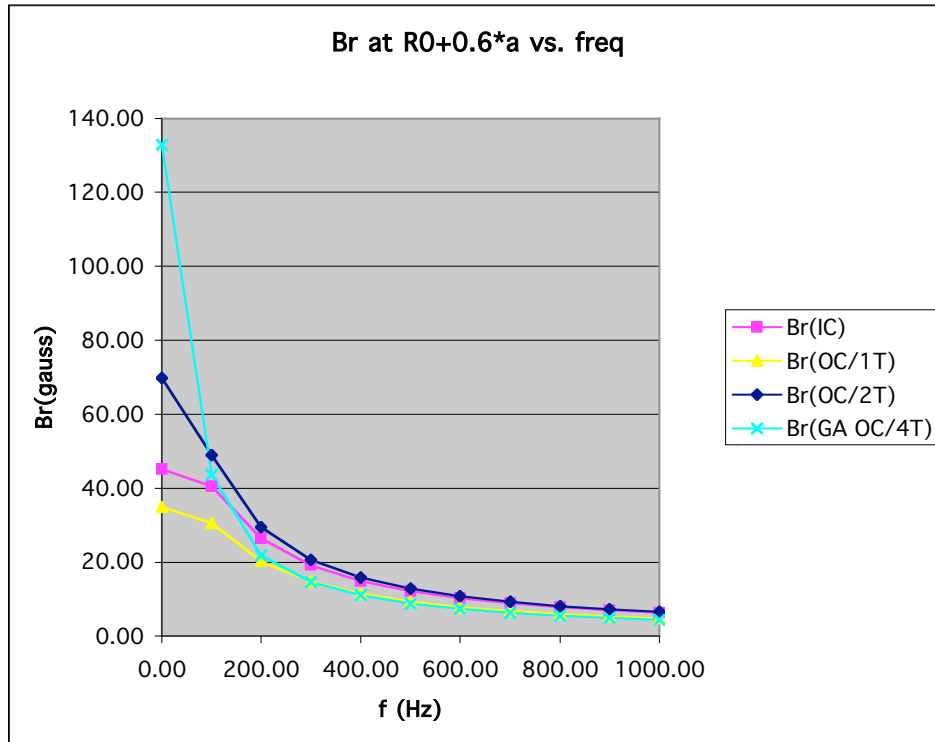


Figure 5 – Radial Field vs. Frequency

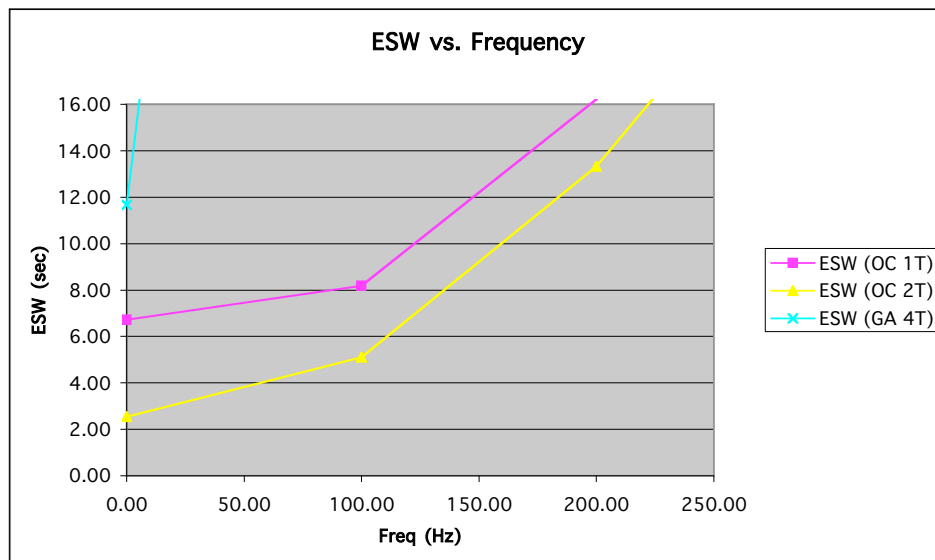


Figure 6 – Available Equivalent Square Wave vs. Frequency for Cable Formed Coils

Based on the analysis performed, the following observations are made.

1 – All of the designs come close to meeting the requirement for 50 gauss under DC conditions. However they all fall short of the requirement for 10 gauss at 1kHz, even with

the simple model used which does not simulate the demagnetizing effect of the eddy currents in the VV. It appears that 10 gauss is only available up to around 500Hz.

2 – The necessity of the additional choke, and higher switching frequency for the new GA internal coils so as to keep the ripple down, appears evident.

3 – At least for the simple model used herein, the ripple requirement dictates the minimum allowable circuit inductance for a given driving voltage. This imposes, therefore, a fundamental limit on system performance. It may actually be advantageous to reduce the base DC voltage (nominally 300 volts) under certain conditions to reduce the ripple.

4 – The effect of the VV is probably very important in determining the ripple response. Considering that the NSTX VV is 5/8" (0.625") thick 304SS, and the skin depth at 1kHz is only 0.550" in 304SS (resistivity =  $7.7 \times 10^{-7}$  ohm-m), the VV may serve as an effective ripple filter, reducing or eliminating the need for external inductance (choke).

5 – If the VV does serve effectively as a ripple filter, then the noise pick-up on in-VV magnetic diagnostics would be reduced in the case of external RWM coils.

6 – For NSTX, it would appear that the resistive and inductive impedance of a cable run from the NTC to FCPC, assumed here to be 250', can be tolerated, allowing the equipment to be installed at FCPC. If closer study finds too small a margin in this regard, perhaps the DC source voltage can be increased above 300V to compensate.

7 – For the cable formed coils (the GA external coil from 4 turns @ 750MCM, the NSTX external coil from 2 turns @ 250MCM, the NSTX coil from 1 turn @ 500MCM) the available pulse length (equivalent square wave) is a function of frequency. For the 500MCM and 750MCM conductors, the allowable pulse length at peak DC current exceeds 5 seconds with a 300 second repetition rate (rated NSTX pulse length and duty factor). However, if a 250MCM cable is used (e.g. to conserve space and allow a 2 turn external coil to be formed) then the pulse length at the low end of the frequency range would be less than 5 seconds as shown in figure 6.

8 – Since the water cooled GA conductor is designed for 7kA continuous then, as long as the DC feeds, etc., are 500MCM or larger, then the RWM could run at rated current for the full 5 second pulse.

9 – The net field available from 3 coil pairs will be higher than that available from one. Figure 7 shows the spatial variation of  $B_r$  (at  $r=R_0+0.6*a$ ) on the midplane for the case of the 1 turn NSTX external coil when all three coil pairs are identically operated. The ratio of multi-coil to single coil field in this case is 2.37.

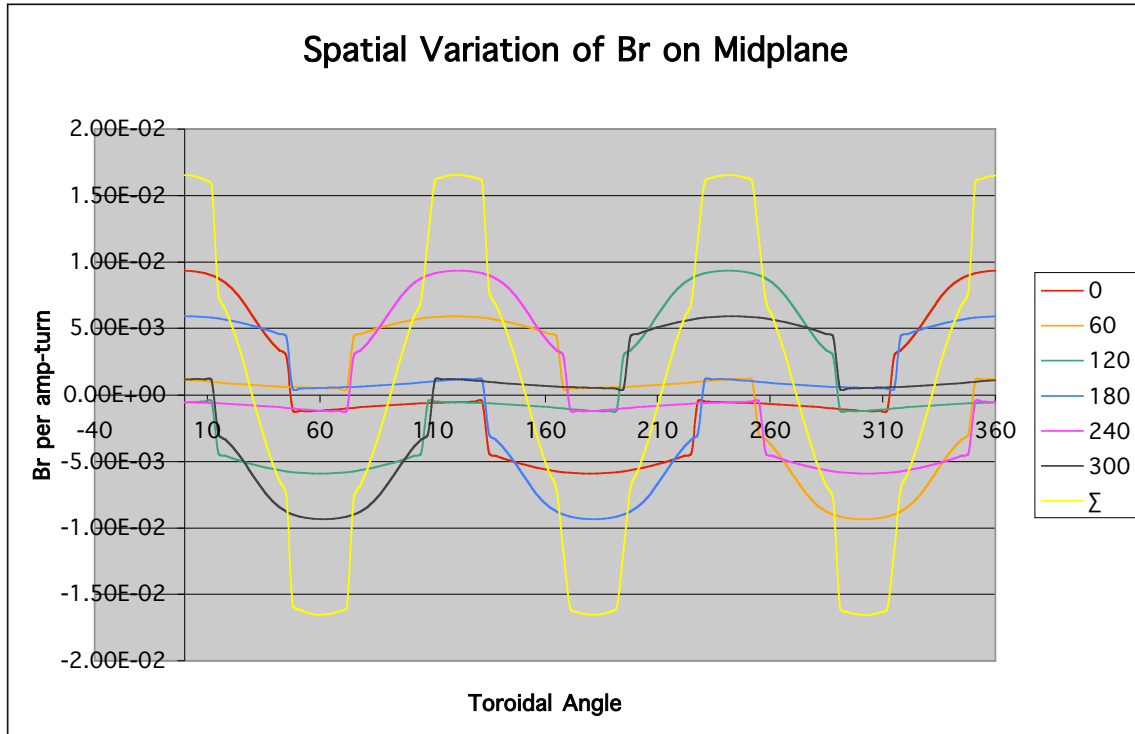


Figure 7 – Spatial Variation of  $B_r$  at  $r=R_0+0.6*a$  on Midplane for 1T External Coil

cc:

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